

Modal Analysis of a Circular Slot Monopole Antenna for UWB application

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Abstract: In this study, a wideband double slot antenna is introduced. The proposed monopole antenna is designed using the theory of characteristic mode for slot monopoles. A circular shape is used as the initial design stage, to enhance the bandwidth, two circular slots are employed. Four modes have been excited to gain a physical insight and to find out which mode is dominant. The design and analysis were completed utilizing both the time domain and multilayer solver in CST 2017, without considering the feeding port. An antenna model in UWB frequency is constructed. Experimented and simulated results shows that the proposed planar structure has a wide impedance width with good radiation characteristics.

Keywords: Characteristic modes (CM), UWB antenna, defected ground structure antenna, monopole antenna.

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1. INTRODUCTION

Designing UWB with concise size and without compromising on antenna performance is a major task nowadays. The UWB systems has become very attractive because of its high transmission data rates and simpler hardware configuration. For wideband operation the circular monopole reception apparatus is a good choice because of the ease of design and construction.

Federal communication commission has dispensed an unlicensed frequency spectrum of 3.1-10.6GHz for UWB frameworks [1]. The patch antenna, which is a critical component of the modern wireless communication system, has played a vital role in the evolution [2]. Microwave radiators compared with traditional microwave reception apparatus are small, ease, easy to produce and they are effectively coordinated into versatile radio and wireless correspondence applications. The fundamental limitation of microstrip antennas are lower gain and impedance width which can be improved by using metamaterials as part of several techniques [3].

For the most part openings are generally cut over a ground plane or over the emanating component to get a slot reception apparatus. Antenna slot with various types of slot shapes have been developed for many specific applications. For example, circular ring [4], fractal [5], arc [6], triangle [7], T- shape [8] and space shapes like ellipse [9] are proposed to acknowledge wide impedance width. To further gain a deep physical insight into the antenna physical properties and radiation mechanisms the theory of characteristic modes (TCM) has been widely

employed in antenna engineering. TCM and its practical applications is presented recently by [10] and [11]. To achieve modal analysis, we have some limited approaches some of which are modal expansion methods [12], and spherical methods [13].

In this letter, modal analysis is developed for a circular monopole reception apparatus. We employed both the time domain solver and multilayer solver in which, the time domain solver is responsible for the S-parameters of the proposed antenna, while the multilayer solver is responsible for the modal analysis of the antenna without considering the feed port. The design and analysis was completed using CST version 2017. To summarize, modal analysis of the proposed structure has been employed to design an antenna suitable for UWB applications. The amplexness of the design strategy is approved through experiment. The remainder of the investigation is sorted out as follows: section 2 introduces the hypothesis behind the modal analysis of antenna, section 3 is the antenna design geometry and parametric analysis of the proposed reception apparatus, section 4 presents the results and discussion of the design and section 5 is the conclusion.

2. THE THEORY OF CHARACTERISTIC MODE

Characteristic mode theory was well described in [14 – 16]. By considering a conducting surface S . The total tangential electric field on the body is given by

$$[L(I) - E^{in}]_{\tan} = 0 \quad (1)$$

Where E^{in} is an incident electric field, and $L(I)$ is the dispersed field due to the current I on S . Introducing the notation

$$Z(I) = [L(I)]_{\text{tan}} \quad (2)$$

Where Z is the symmetric operator, and setting $Z = R + jX$, the following weighted eigenvalue problem is defined:

$$Z(I_n) = V_n R(I_n) \quad (3)$$

V_n is the eigenvalues, the I_n is the modal currents and R is a weight operator. By using $V_n = 1 + j\lambda_n$ and cancel the common term $R(I_n)$ in (3) we then obtain

$$X(I_n) = \lambda_n R(I_n) \quad (4)$$

The mode currents from a weighted orthogonal set over the conductor's surface S . Each modal current, which can be normalized according to radiated unit power.

$$\langle I_n^*, R I_n \rangle = 1 \quad (5)$$

The current I , on a conducting body can be expanded as

$$I = \sum_{n=1}^N \alpha_n I_n \quad (6)$$

Where the I_n is the basic functions of the mode current and the λ_n is the modal expanded coefficients. For an electrically small radiator, only a few modes are required to characterized the antennas behavior. Modes with little λ_n are good radiators, those with huge λ_n are poor radiators, while modes with $\lambda_n = 0$ are at resonance [17] and [18]. The extend of eigenvalues λ_n thus tells how well a particular mode radiates. The modal significance (MS) is defined as

$$MS = \left| \frac{1}{1 + j\lambda_n} \right| \quad (7)$$

Which is related to the conducting body's shape, dimension and frequency, but is independent of any specific excitation. Characteristic angle is defined in [19] and [20] as

$$\theta = 180 - \tan^{-1}(\lambda_n) \quad (8)$$

Equation (8) gives a clear elucidation of how energy is stored in the monopole antenna, when the angle is 180° the mode resonates, when it is less than 180° the mode stores magnetic energy and when it is greater than 180° the mode stores electric energy.

3. DESIGN STRATEGY

3.1 Antenna Geometry

The geometry of UWB circular reception apparatus is pictured in Figure 1. The receiving wire is simulated utilizing both time domain and multilayer solver analysis in CST version 2017. The system is deployed on a dielectric substrate: FR4 epoxy with an overall

permittivity of 4.3, 0.025 loss tangent and stature of 1.6mm. The absolute structure zone was to improve the impedance width of the proposed receiving wire from 2.8GHz to 12GHz, two slots are set at each side of the transmitting component with a defected ground plane. The two openings assume an imperative job in improving the impedance width of the proposed reception apparatus. Great portrayal as far as wideband and radiation properties are the benefits of circular shaped reception apparatus. The enhanced elements of the circular formed receiving wire are delineated in Table 1 below. Another structure which we can called it a defective structure has also plays a very important role for improving the performance of the monopole antenna, as clearly illustrated in figure 1 b. The location of the gap controls the bandwidth of the monopole antenna for the fact that, current distribution in the ground will create influences and the behavior of the receiving wire.

The modal analysis was applied to the monopole reception apparatus without feeding port using multilayer solver which represents the standardized amplitude of the modal current. This standardized amplitude doesn't rely upon the feeding port however just relies upon the shape, and size of the conducting object. Another important factor to be consider is the use of loss free substrate instead of lossy parameters because loss free is the prerequisite of using multilayer solver in CST.

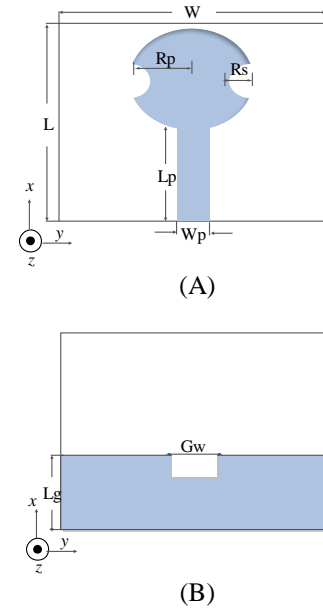


Figure 1. Circular monopole antenna geometry: A, top layer, B, ground plane

Table 1. Dimension parameters of the proposed antenna

Parameter	Size (mm)	Parameter	Size (mm)
L	30	R_S	2.6
L_g	10	W	30
L_p	11.05	W_p	3
R_p	9.35	G_w	3

3.2 Parametric analysis of the design

The parametric analysis is presented using different values of R_s , that is the radius of the slots from Figure 2, in order to have an optimum value that can give us a better performance of the reception apparatus regarding the reflection coefficient, radiation characteristics and bandwidth. Figure 3.6 shows S_{11} for the various radius of the slot as can be seen from the plot, when the value of R_s is 2.6mm the resulting S_{11} is perfect at that value. Therefore, we used 2.6mm as the radius of the slots for both the two reception apparatus. Even though, the remaining plots are also within the range of -10dB, also has a good result but with 2.6mm as the value of radius the performance has improved that's is why the value is taken as the final value for the design and analysis of the proposed reception apparatus.

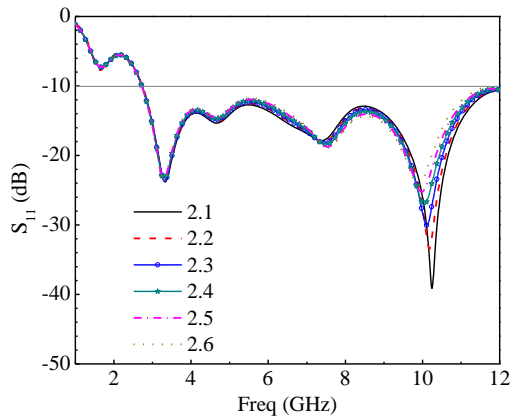


Figure 2. S_{11} plot for various radius R_s (mm)

4. RESULTS AND DISCUSSION

In light of the structure parameters appeared in Figure 1, the single reception apparatus was simulated and analyzed. The construction of the single element and measurement were carried out at the Institute of Micro-Nano Photo-electronic and Electromagnetic Technology Innovation School of Electronics and Information Engineering, Hebei University of technology, China using N5244A network analyzer. Figure 3 presents the measured and simulated $|S_{11}|$ against frequency of the circular monopole antenna. It can be observed from Figure 3 that the $|S_{11}|$ is < -10 dB with an impedance percentage bandwidth of 200% from 2.8GHz to 12GHz. The constructed photo of single element circular shaped antenna is pictured in Figure 4.

The modal significance of the first four modes at 3.35GHz is presented in Figure 5, only mode 1 has exactly modal significance of up to 1.0. The remaining mode has been shifted a little bit, therefore, only mode 1 will dominate at this frequency response. At 5.5GHz mode 1 has exactly modal significance of nearly 1.0, mode 2, 3 and 4 is not at 5.5GHz, therefore mode 1 will also dominate at this frequency as shown in Figure 6. The validity of the model further implies that we can consider the antenna shape and feed design as independent steps. Figure 7 presents the characteristic angle for the circular monopole antenna at mode 1, 2, 3 and 4. As mentioned earlier a mode resonates when $\lambda_n = 0$, that is when the angle is 180° . From the figure mode 1 is at 3.35GHz and

remains the frequency of interest. At 5.5GHz it is only mode 1 that is at that frequency as shown in Figure 9, the remaining modes have been shifted. It was clearly shown that for all the two frequency of interest the characteristic angle is at 180° which signifies that the mode is a good radiator.

Figure 10 and Figure 11 presents the eigenvalues for the four modes at 3.35GHz and 5.5GHz. From the results it clearly shows that the modes are at resonant with $\lambda_n = 0$ but mode 1 is more dominant. Figure 12 is the normalized 2D modal far field pattern at 3.35GHz for four modes while figure 13 remain the normalized 2D pattern at 5.5GHz for mode 1, 2, 3 and 4 respectively. The pattern shows the XZ-plane and YZ-planes at the frequency of interest.

Figures 13 to 16 presents the modal surface current distribution and the modal 3D far field pattern of four different modes at 3.35GHz

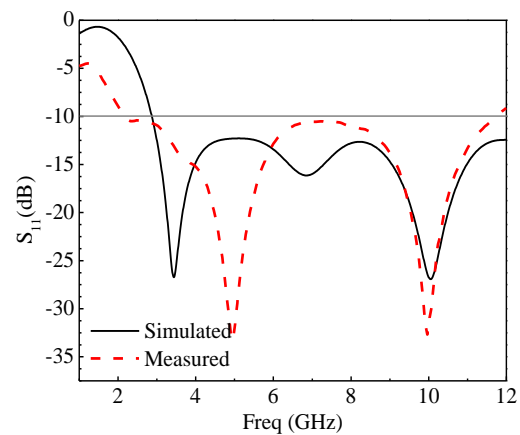
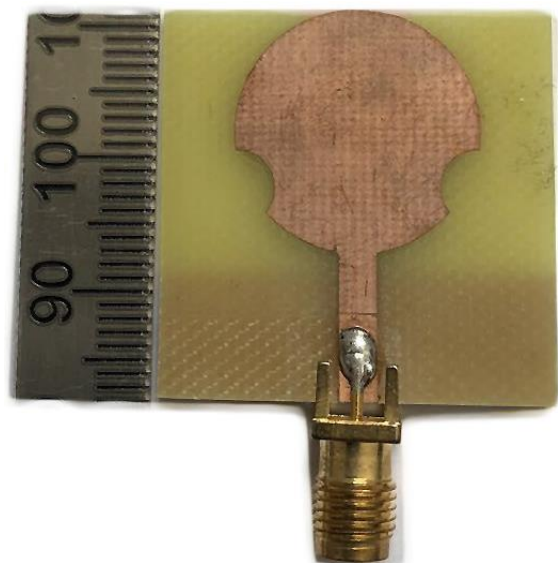
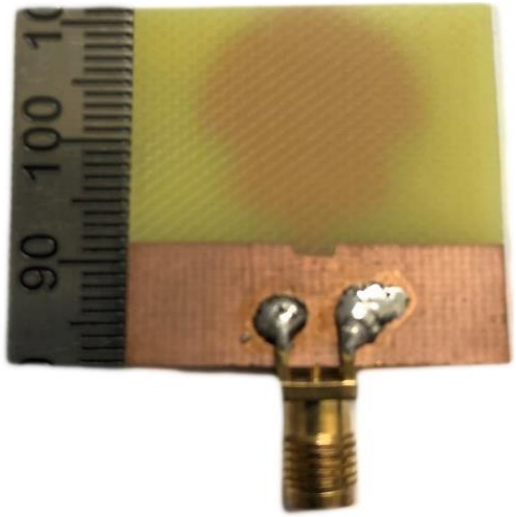


Figure 3. Measured and simulated S_{11} monopole antenna



(A)



(B)

Figure 4. Fabricated prototype: A, top view, B, bottom view

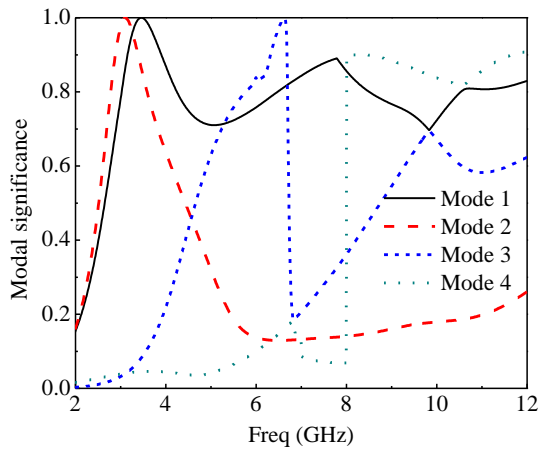


Figure 5. Modal significance at 3.35GHz

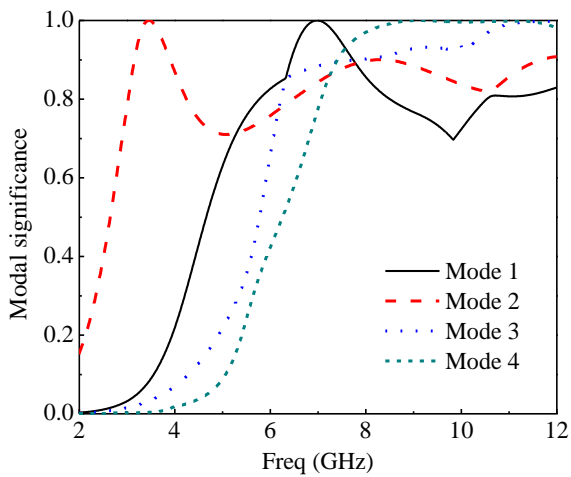


Figure 6. Modal significance at 5.5GHz

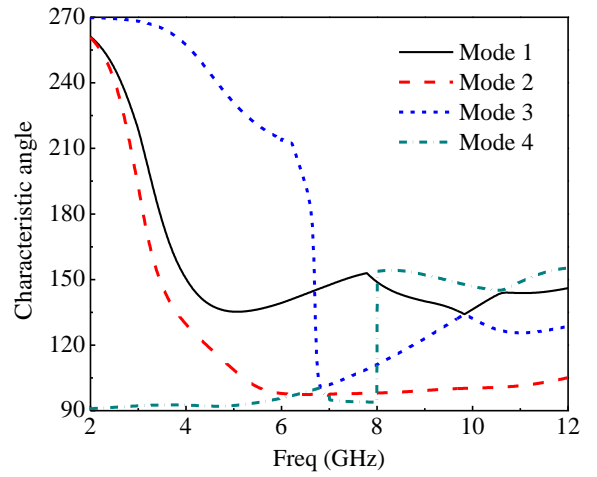


Figure 7. Characteristic angle at 3.35GHz

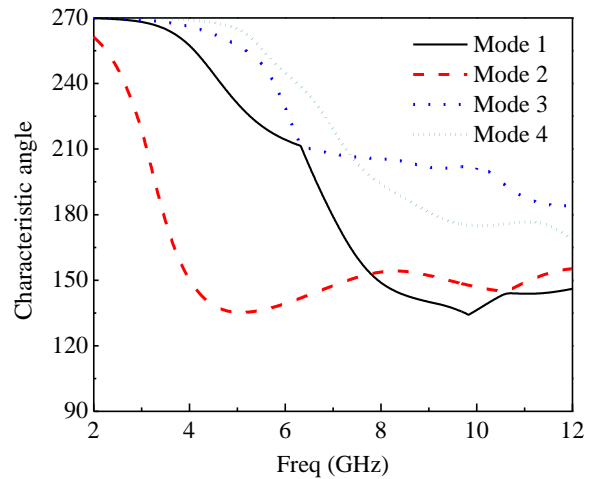


Figure 8. Characteristic angle at 5.5GHz

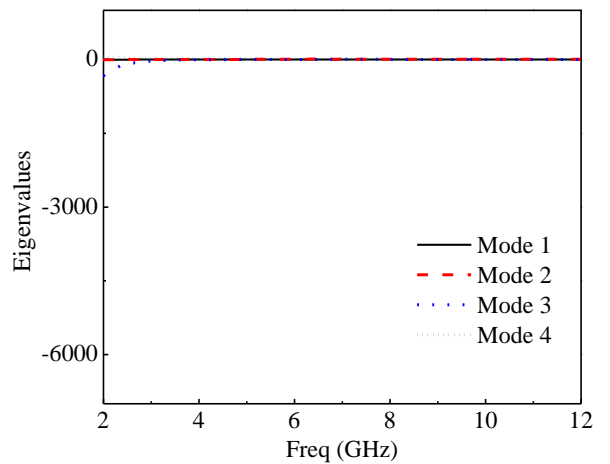


Figure 9. Eigenvalues at 3.35GHz

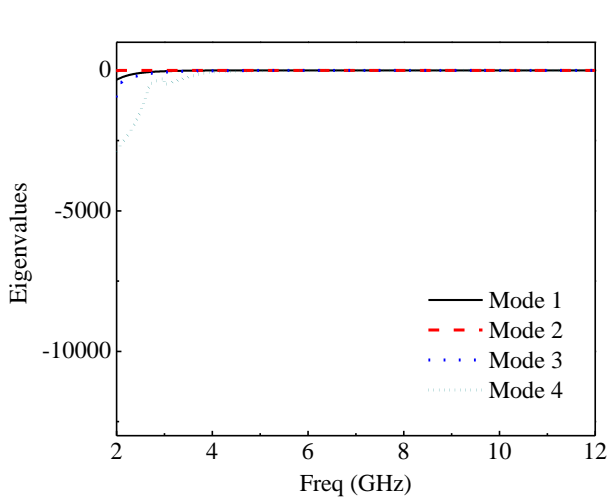
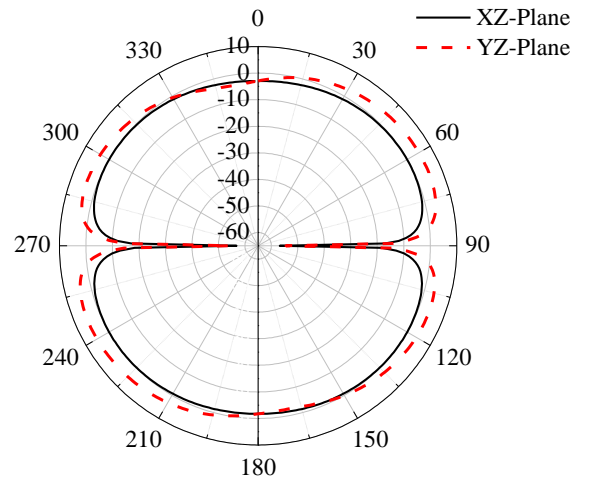
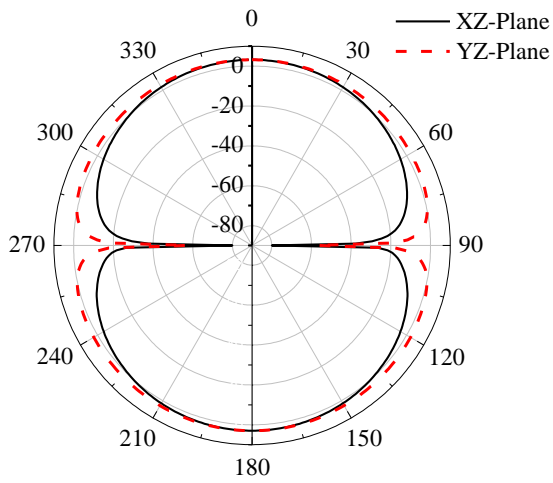


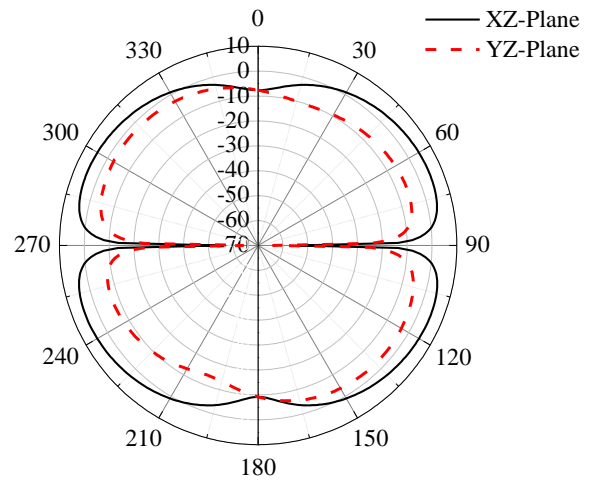
Figure 10. Eigenvalues at 5.5GHz



(c)

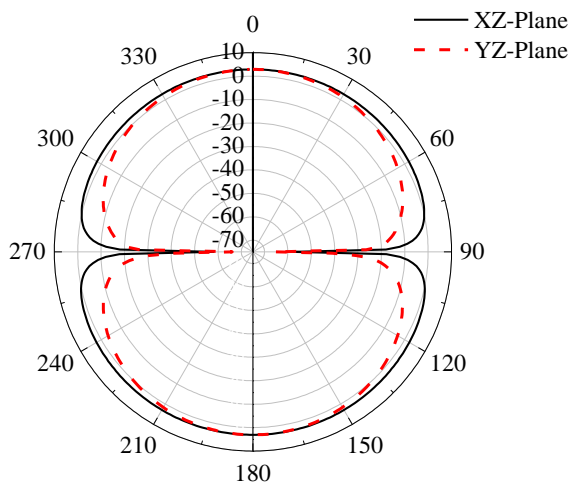


(a)

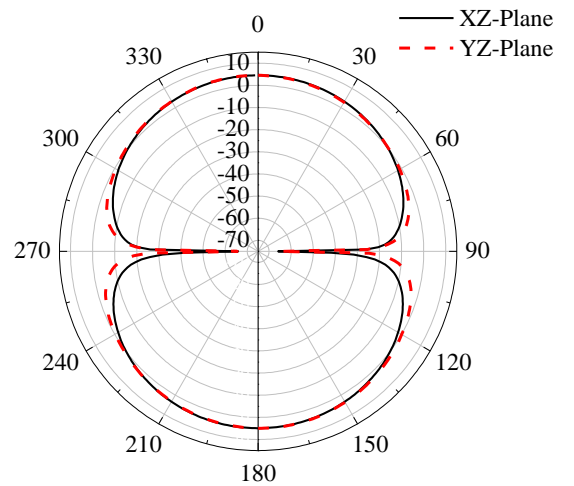


(d)

Figure 11. Far field pattern at 3.35GHz for (a) mode 1, (b) mode 2, (c) mode 3 and (d) mode 4



(b)



(a)

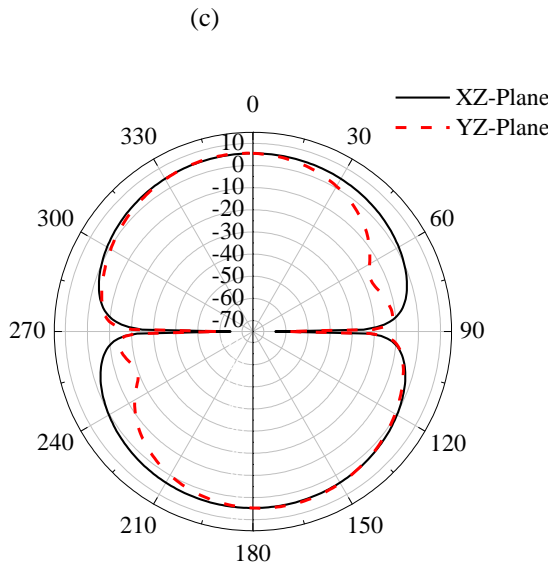
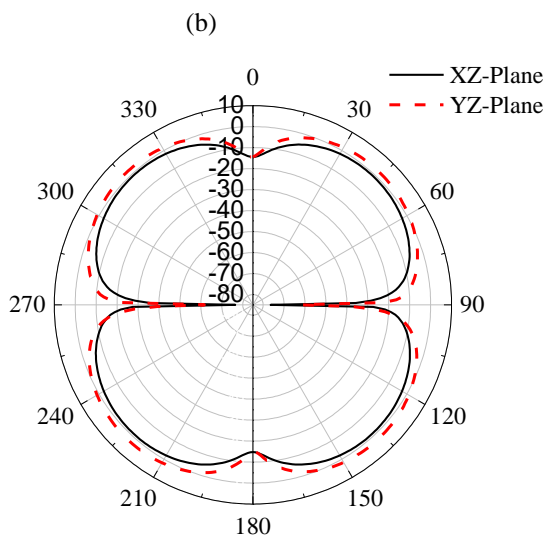
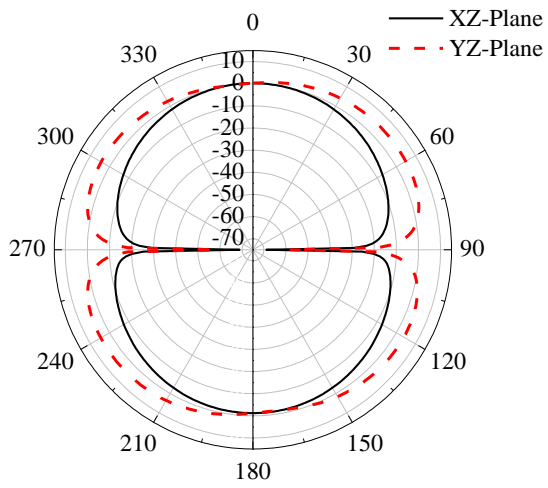
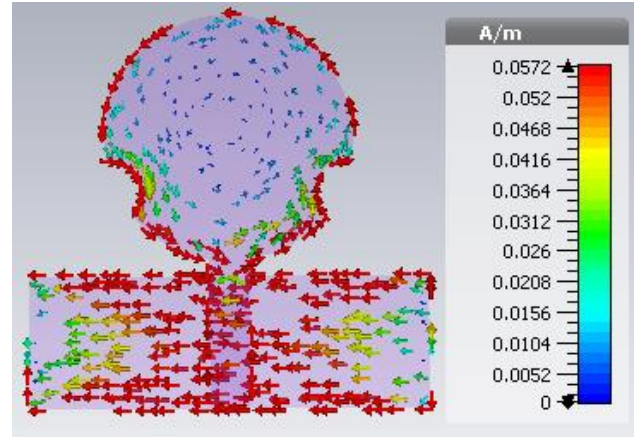
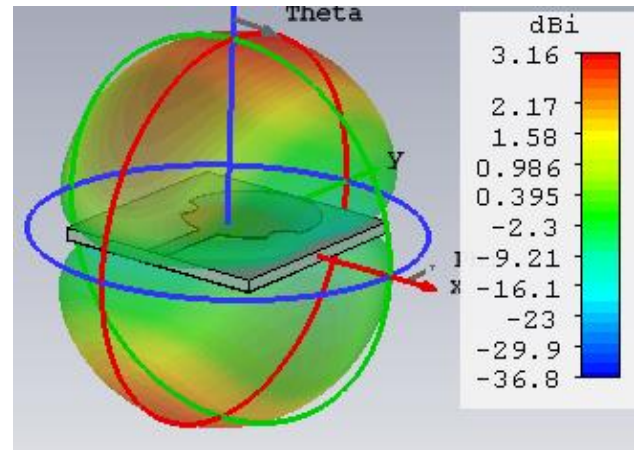


Figure 12. Far field pattern at 5.5GHz for (a) mode 1, (b) mode 2, (c) mode 3 and (d) mode 4

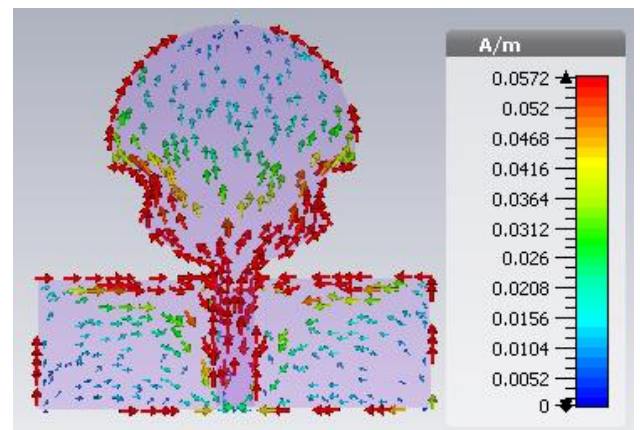


(A)

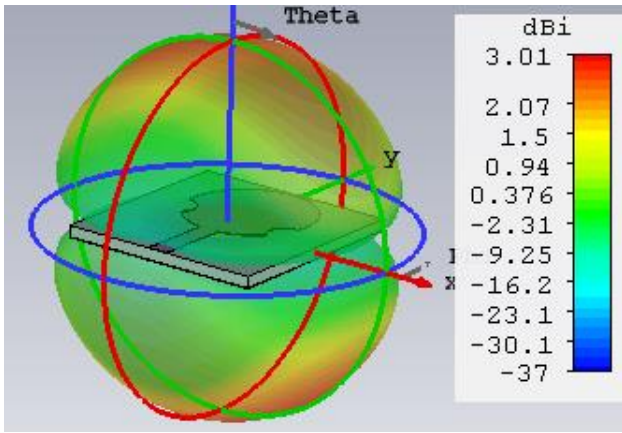


(B)

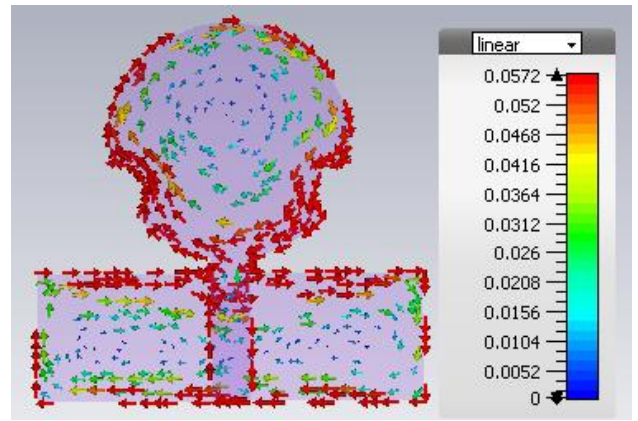
Figure 13. Modal surface current (A), 3D radiation pattern of mode 1 at 3.35GHz (B).



(A)

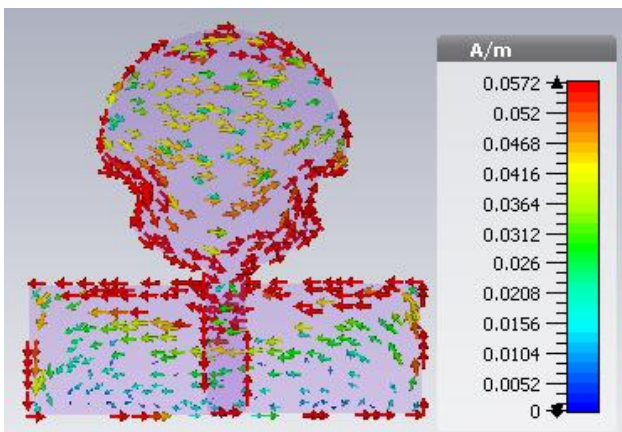


(B)

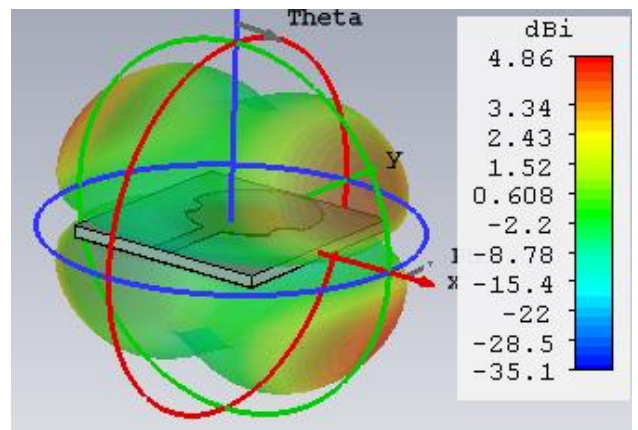


(A)

Figure 14. Modal surface current (A), 3D radiation pattern of mode 2 at 3.35GHz (B).

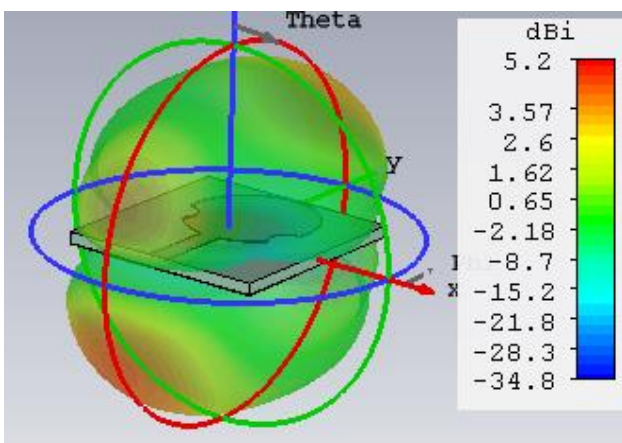


(A)



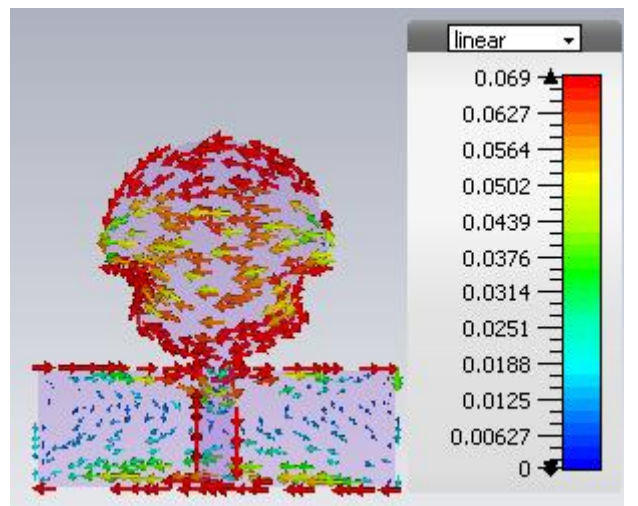
(B)

Figure 16. Modal surface current (A), 3D radiation pattern of mode 4 at 3.35GHz (B).



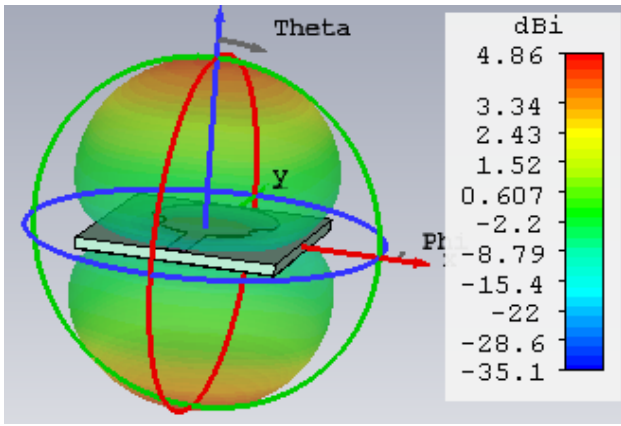
(B)

Figure 15. Modal surface current (A), 3D radiation pattern of mode 3 at 3.35GHz (B).

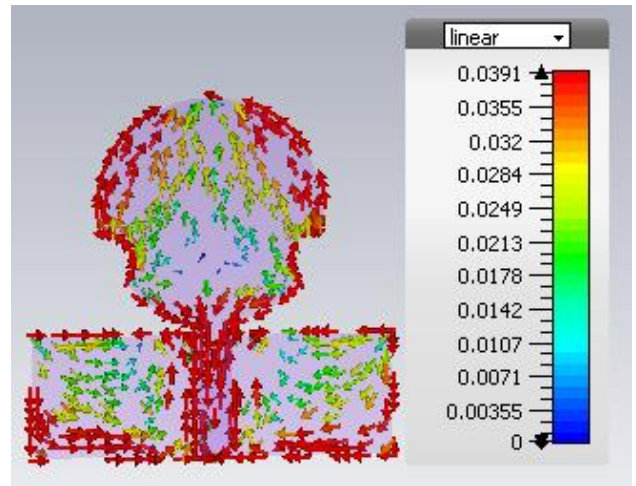


(A)

Figures 17 to 20 presents the modal surface current distribution and the modal 3D far field pattern of four different modes at 5.5GHz. As can be clearly seen It is obvious that the structure can support more than one mode in the frequency band of interest in two orthogonal directions.

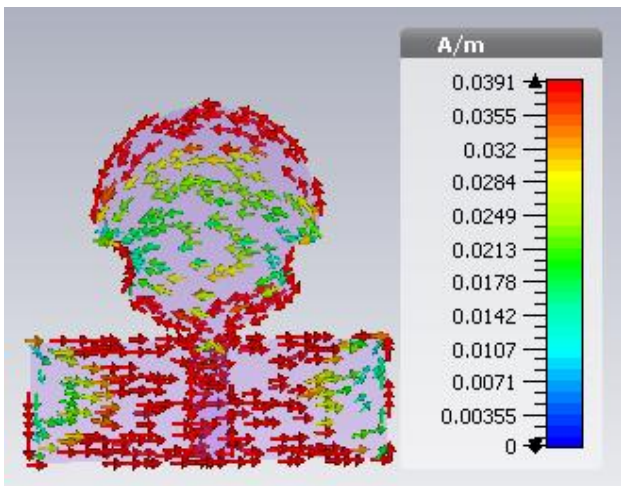


(B)

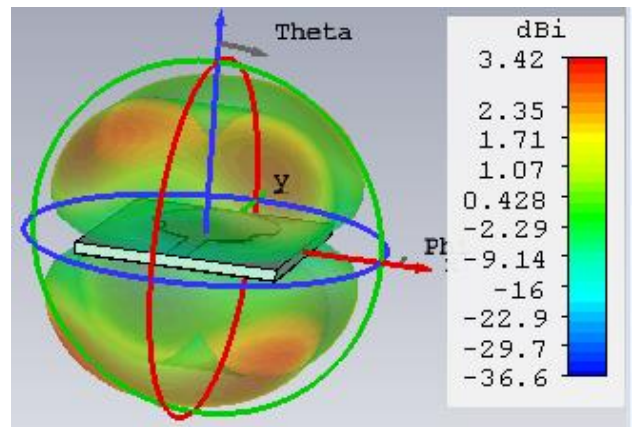


(A)

Figure 17. Modal surface current (A), 3D radiation pattern of mode 1 at 5.5GHz (B).

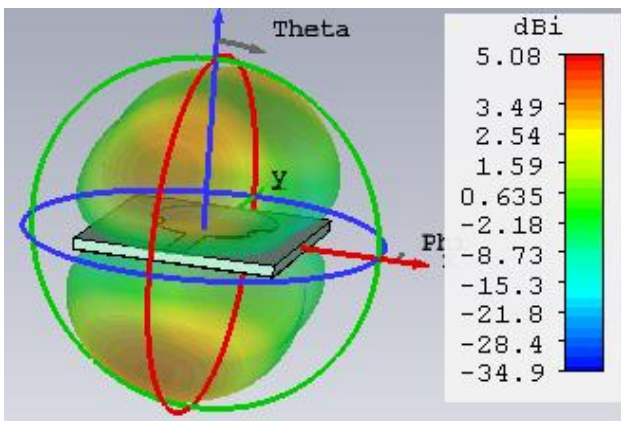


(A)

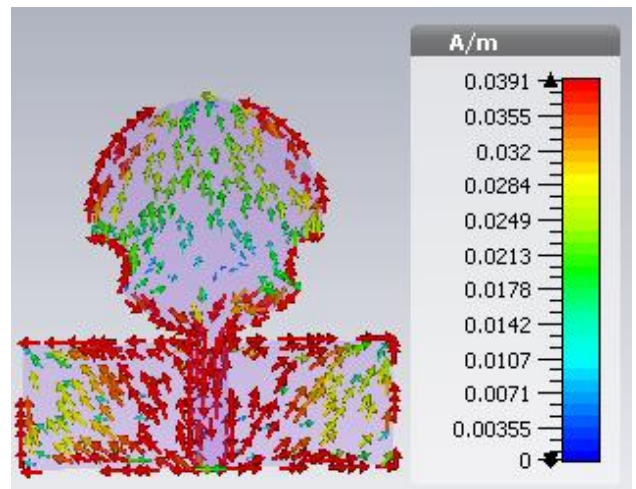


(B)

Figure 19. Modal surface current (A), 3D radiation pattern of mode 3 at 5.5GHz (B).

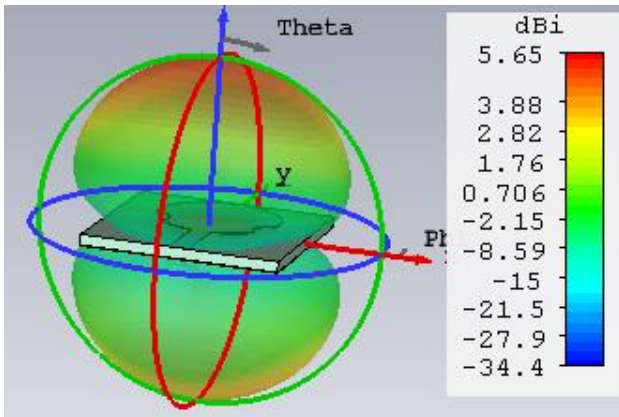


(B)



(A)

Figure 18. Modal surface current (A), 3D radiation pattern of mode 2 at 5.5GHz (B).



(B)

Figure 20. Modal surface current (A), 3D radiation pattern of mode 4 at 5.5GHz (B).

The bandwidth of the proposed receiving wire is contrasted with past works in Table 2. The proposed circular monopole antenna achieves a considerable wide impedance width and smaller size as compared with tri-notched UWB antenna [4], fractal UWB MIMO antenna [5] and circularly polarized slot antenna [6].

Table 2. Comparison between the reference antenna

Reference	Bandwidth (GHz)	Size (mm)
[4]	3.1 - 11	34x36x1
[5]	3.1 - 12.5	24x32x0.8
[6]	2.1 – 5.6	42x42x1.5
[7]	2.9 – 19.2	40x40x1.5
Prop.	2.6 - 12	30x30x1.6

3. CONCLUSION

The CMA has been applied to a circular monopole antenna using multilayer solver analysis in CST version 2017. The modal significance clearly shows that mode 1 is the dominant mode. The design procedure also removes the additional burden that may arise as a result of feeding port. A prototype has been constructed. S-parameters have been measured to validate the design. Owing to the wide BW, the proposed antenna finds wide application in UWB communication systems.

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